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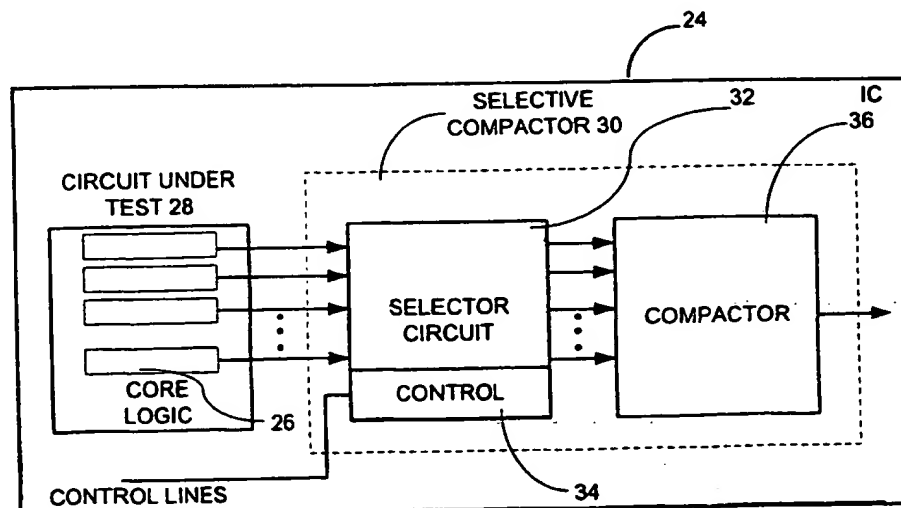
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(54) Title: METHOD AND APPARATUS FOR SELECTIVELY COMPACTING TEST RESPONSES



(57) Abstract: A method and apparatus to compact test responses containing unknown values or multiple fault effects in a deterministic test environment. The proposed selective compactor (30) employs a linear compactor (36) with selection circuitry (32) for selectively passing test responses to the compactor (36). In one embodiment, gating logic is controlled by a control register, a decoder, and flag registers. This circuitry, in conjunction with any conventional parallel test-response compaction scheme, allows control circuitry (34) to selectively enable serial outputs of desired scan chains to be fed into a parallel compactor (36) at a particular clock rate. A first flag register determines whether all, or only some, scan chain outputs are enabled and fed through the compactor (36). A second flag register determines if the scan chain selected by the selector register is enabled and all other scan chains are disabled, or the selected scan chain is disabled and all other scan chains are enabled.



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**METHOD AND APPARATUS FOR SELECTIVELY
COMPACTING TEST RESPONSES**

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TECHNICAL FIELD

This invention relates generally to testing of integrated circuits and more particularly relates to compaction of test responses used in testing for faults in integrated circuits.

10

BACKGROUND

As integrated circuits are produced with greater and greater levels of circuit density, efficient testing schemes that guarantee very high fault coverage while minimizing test costs and chip area overhead have become essential. However, as the complexity of circuits continues to increase, high fault coverage of several types of fault models becomes more difficult to achieve with traditional testing paradigms. This difficulty arises for several reasons. First, larger integrated circuits have a very high and still increasing logic-to-pin ratio that creates a test data transfer bottleneck at the chip pins. Second, larger circuits require a prohibitively large volume of test data that must be then stored in external testing equipment. Third, applying the test data to a large circuit requires an increasingly long test application time. And fourth, present external testing equipment is unable to test such larger circuits at their speed of operation.

Integrated circuits are presently tested using a number of structured design for testability (DFT) techniques. These techniques rest on the general concept of making all or some state variables (memory elements like flip-flops and latches) directly controllable and observable. If this can be arranged, a circuit can be treated, as far as testing of combinational faults is concerned, as a combinational or a nearly combinational network. The most-often used DFT methodology is based on scan chains. It assumes that during testing all (or almost all) memory elements are connected

into one or more shift registers, as shown in United States Patent No. 4,503,537. A circuit that has been designed for test has two modes of operation: a normal mode and a test, or scan, mode. In the normal mode, the memory elements perform their regular functions. In the scan mode, the memory elements become scan cells that are connected
5 to form a number of shift registers called scan chains. These scan chains are used to shift a set of test patterns into the circuit and to shift out circuit, or test, responses to the test patterns. The test responses are then compared to fault-free responses to determine if the circuit under test (CUT) works properly.

Scan design methodology has gained widespread adoption by virtue of its
10 simple automatic test pattern generation (ATPG) and silicon debugging capabilities. Today, ATPG software tools are so efficient that it is possible to generate test sets (a collection of test patterns) that guarantee almost complete fault coverage of several types of fault models including stuck-at, transition, path delay faults, and bridging faults. Typically, when a particular potential fault in a circuit is targeted by an ATPG
15 tool, only a small number of scan cells, e.g., 2-5%, must be specified to detect the particular fault (deterministically specified cells). The remaining scan cells in the scan chains are filled with random binary values (randomly specified cells). This way the pattern is fully specified, more likely to detect some additional faults, and can be stored on a tester.

20 FIG. 1 is a block diagram of a conventional system 10 for testing digital circuits with scan chains. External automatic testing equipment (ATE), or tester, 12 applies a set of fully specified test patterns 14 one by one to a CUT 16 in scan mode via scan chains 18 within the circuit. The circuit is then run in normal mode using the test pattern as input, and the test response to the test pattern is stored in the scan chains. With the
25 circuit again in scan mode, the response is then routed to the tester 12, which compares the response with a fault-free reference response 20, also one by one. For large circuits, this approach becomes infeasible because of large test set sizes and long test application times. It has been reported that the volume of test data can exceed one kilobit per single

logic gate in a large design. The significant limitation of this approach is that it requires an expensive, memory-intensive tester and a long test time to test a complex circuit.

These limitations of time and storage can be overcome to some extent by adopting a built-in self-test (BIST) framework as shown in FIG. 2. In BIST, additional on-chip circuitry is included to generate test patterns, evaluate test responses, and control the test. For example, a pseudo-random pattern generator 21 is used to generate the test patterns, instead of having deterministic test patterns. Additionally, a multiple input signature register (MISR) 22 is used to generate and store a resulting signature from test responses. In conventional logic BIST, where pseudo-random patterns are used as test patterns, 95-96% coverage of stuck-at faults can be achieved provided that test points are employed to address random-pattern resistant faults. On average, one to two test points may be required for every 1000 gates. In BIST, all responses propagating to observable outputs and the signature register have to be known. Unknown values corrupt the signature and therefore must be bounded by additional test logic. Even though pseudo-random test patterns appear to cover a significant percentage of stuck-at faults, these patterns must be supplemented by deterministic patterns that target the remaining, random pattern resistant faults. Very often the tester memory required to store the supplemental patterns in BIST exceeds 50% of the memory required in the deterministic approach described above. Another limitation of BIST is that other types of faults, such as transition or path delay faults, are not handled efficiently by pseudo-random patterns. Because of the complexity of the circuits and the limitations inherent in BIST, it is extremely difficult, if not impossible, to provide a set of test patterns that fully covers hard-to-test faults.

Some of the DFT techniques include compactors to compress the test responses from the scan chains. There are generally two types of compactors: time compactors and spatial compactors. Time compactors typically have a feedback structure with memory elements for storing a signature, which represents the results of the test. After the signature is completed it is read and compared to a fault-free signature to determine if an error exists in the integrated circuit. Spatial compactors generally compress a

collection of bits (called a vector) from scan chains. The compacted output is analyzed in real time as the test responses are shifted out of the scan chains. Spatial compactors can be customized for a given circuit under test to reduce the aliasing phenomenon, as shown in the United States Patent 5,790,562 and in few other works based on
5 multiplexed parity trees or nonlinear trees comprising elementary gates such as AND, OR, NAND, and NOR gates.

Linear spatial compactors are built of Exclusive-OR (XOR) or Exclusive-NOR (XNOR) gates to generate n test outputs from the m primary outputs of the circuit under test, where $n < m$. Linear compactors differ from nonlinear compactors in that the
10 output value of a linear compactor changes with a change in just one input to the compactor. With nonlinear compactors, a change in an input value may go undetected at the output of the compactor. However, even linear compactors may mask errors in an integrated circuit. For example, the basic characteristic an XOR (parity) tree is that any combination of odd number of errors on its inputs propagates to their outputs, and any
15 combination of even number of errors remains undetected.

An ideal compaction algorithm has the following features: (1) it is easy to implement as a part of the on-chip test circuitry, (2) it is not a limiting factor with respect to test time, (3) it provides a logarithmic compression of the test data, and (4) it does not lose information concerning faults. In general, however, there is no known
20 compaction algorithm that satisfies all the above criteria. In particular, it is difficult to ensure that the compressed output obtained from a faulty circuit is not the same as that of the fault-free circuit. This phenomenon is often referred to as error masking or aliasing and is measured in terms of the likelihood of its occurrence. An example of error masking occurs when the spatial compactor reads two fault effects at the same
25 time. The multiple fault effects cancel each other out and the compactor output is the same as if no faults occurred.

Unknown states are also problematic for error detection. An unknown state on one or more inputs of an XOR tree generates unknown values on its output, and

consequently masks propagation of faults on other inputs. A common application of space compactors is to combine the observation points inserted into the CUT as a part of design-for-testability methodology. The spatial compactors can be also used to reduce the size of the time compactors by limiting the number of their parallel inputs.

- 5 Undoubtedly, the most popular time compactors used in practice are linear feedback shift registers (LFSRs). In its basic form, the LFSR (see FIG. 3) is modified to accept an external input in order to act as a polynomial divider. An alternative implementation (called type II LFSR) is shown in FIG. 4. The input sequence, represented by a polynomial, is divided by the characteristic polynomial of the LFSR.
- 10 As the division proceeds, the quotient sequence appears at the output of the LFSR and the remainder is kept in the LFSR. Once testing is completed, the content of the LFSR can be treated as a signature.

- FIG. 5 shows another time compactor (which is a natural extension of the LFSR-based compactor) called a multiple-input LFSR, also known as a multiple-input
- 15 signature register (MISR). The MISR is used to test circuits in the multiple scan chain environment such as shown in the United States Patent 4,503,537. MISRs feature a number of XOR gates added to the flip-flops. The CUT scan chain outputs are then connected to these gates.

- FIG. 6 shows an example of a pipelined spatial compactor with a bank of flip-
- 20 flops separating stages of XOR gates. A clock (not shown) controls the flip-flops and allows a one-cycle delay before reading the compacted output.

- The limitation of spatial compactors, such as the one shown in FIG. 6, is that unknown states can reduce fault coverage. Time compactors, such as shown in FIGS. 3,
- 4, and 5, are completely unable to handle unknown states since an unknown state on
- 25 any input can corrupt the compressed output generated by the compactor. With both time compactors and spatial compactors, multiple fault effects can reduce fault

coverage. Additionally, if a fault effect is detected within the integrated circuit, these compactors have limited ability to localize the fault.

An object of the invention, therefore, is to provide an efficient compactor that can select which scan chains are analyzed. This ability to select allows the compactor to generate a valid compressed output even when receiving unknown states or multiple fault effects on its inputs. The compactor can also be used diagnostically to determine the location of faults within an integrated circuit.

SUMMARY

A compactor is disclosed that selects test responses in one or more scan chains to compact into a compressed output, while one or more other test responses are masked. Thus, test responses containing unknown states may be masked to ensure that the compactor generates a valid compressed output. Additionally, test responses can be masked to ensure fault masking does not occur. The compactor can also analyze test responses from individual scan chains to diagnostically localize faults in an integrated circuit.

A compactor includes selection circuitry that controls which scan chains are analyzed. The selection circuitry passes desired test responses from scan chains onto a compactor, while masking other test responses. In one embodiment, the selection circuitry may include an identification register that is loaded with a unique identifier of a scan chain. Based on the state of a flag register, either only the test response stored within the scan chain identified is passed to the compactor or all test responses are passed to the compactor except the test response associated with the identified scan chain.

In another embodiment, the selection circuitry includes a flag that controls whether only selected test responses are compacted or whether all test responses are compacted.

In yet another embodiment, a control register is used that individually identifies each scan chain included in compaction. In this embodiment, a variable number (e.g., 1, 2, 3, 4...) of test responses within scan chains may be included in compaction.

Alternatively, the control register may store a unique identifier that is decoded to select one test response that is compacted.

In still another embodiment, the selection circuitry includes a control line that masks bits from scan chains on a per clock-cycle basis. Consequently, a test response may have only individual bits masked while the remaining bits of the test response are compacted.

The foregoing and other objects, features, and advantages of the invention will become more apparent from the following detailed description, which proceeds with reference to the following drawings.

BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1 is a block diagram of a prior art system for testing integrated circuits.

FIG. 2 is a block diagram of a prior art system using a built-in-test system.

FIG. 3 is a circuit diagram of a prior art type I LFSR compactor.

FIG. 4 is a circuit diagram of a prior art type II LFSR compactor.

FIG. 5 is a circuit diagram of a prior art architecture of a multiple input signature register (MISR) compactor shown receiving input from scan chains.

FIG. 6 is a circuit diagram of a prior art pipelined spatial compactor.

FIG. 7 is a block diagram of a selective compactor according to the invention.

FIG. 8 shows one embodiment of a selective compactor, including selection circuitry and a spatial compactor, for masking test responses from scan chains.

FIG. 9 is another embodiment of a selective compactor including selection circuitry and a time compactor for masking test responses from scan chains.

the update line deactivated. When the update line is activated, the multiplexer passes the contents of register 52 to corresponding bit positions of the control register 64. The control register 64 is then loaded synchronously with the clock.

The selector circuit 46 includes logic gates, shown generally at 68, coupled to
5 the control register 64. The logic gates 68 are responsive to the different fields 58, 60, 62 of the control register 64. For example, the scan identification field 58 contains a sufficient number of bits to uniquely identify any of the scan chains 44. The scan identification field 58 of the control register 64 is connected to a decoder, shown at 70 as a series of AND gates and inverters. The decoder 70 provides a logic one on a
10 decoder output depending on the scan identification field, while the other outputs of the decoder are a logic zero.

The one/not one field 60 of the control register 64 is used to either pass only one test response associated with the scan chain identified in the scan identification field 58 or pass all of the test responses except for the scan chain identified in the scan
15 identification field. The all/not all field 62 is effectively an override of the other fields. In particular, field 62 controls whether all of the test responses in the scan chains 44 are passed to the compactor 48 or only the test responses as controlled by the scan identification field 58 and the one/not one field 60. With field 62 cleared, only test responses as controlled by the scan identification field 58 and field 60 pass to the
20 compactor 48. Conversely, if the field 62 is set to a logic one, then all of the test responses from all of the scan chains 44 pass to the compactor 48 regardless of the scan identification field 58 and the one/not one field 60.

FIG. 9 shows another embodiment of a selective compactor 80 that is coupled to scan chains 82. The selective compactor includes a selector circuit 84, which is identical
25 to the selector circuit 46 described in relation to FIG. 8. The selective compactor 80 also includes a time compactor 84, which is well understood in the art to be a circular compactor. The time compactor includes multiple flip-flops 86 and XOR gates 88 coupled in series. A reset line 90 is coupled to the flip-flops 86 to reset the compactor 84. The reset line may be reset multiple times while reading the scan chains. Output

register 92 provides a valid output of the compactor 84 upon activation of a read line 94.

Referring to both FIGs. 8 and 9, in operation the scan chains 82 are serially loaded with predetermined test patterns by shifting data on scan channels (not shown) from an ATE (not shown). Simultaneously, the input register 52 is loaded with a scan identification and the controlling flags in fields 60, 62. The test patterns in the scan chains 44, 82 are applied to the circuit under test and test responses are stored in the scan chains. Prior to shifting the test responses out of the scan chains, the update line 66 is activated, thus moving fields 58, 60, 62 to the control register 64. The control register thereby controls the logic gates 68 to select the test responses that are passed to the compactors 48, 84. If the field 62 is in a state such that selection is not overridden, then certain of the test responses are masked. In the example of FIG. 8, the spatial compactor 48 provides the corresponding compressed output serially and simultaneously with shifting the test responses out of the scan chains. Conversely, in FIG. 9 the selective compactor 80 does not provide the appropriate compressed output until the read line 94 is activated. The selective compactor 80 provides a parallel compressed output as opposed to serial. The selective compactor 80 may be read multiple times (e.g., every eighth clock cycle) while reading out the test responses.

FIG. 10 shows another embodiment of a selective compactor 100. Again, the selective compactor includes a selector circuit 102 and a compactor 104. The compactor 104 is a type of spatial compactor called a cascaded compactor. N scan chains 106 include M scan cells 108, each of which stores one bit of the test response. The selector circuit 102 includes logic gates 110, in this case shown as AND gates, coupled to a control line 112. The compactor 104 is a time compactor with a single serial output 114. The control line 112 is used to mask the test responses. In particular, the control line 112 either masks all corresponding scan cells in the scan chains or allows all of the scan cells to pass to the compactor 80. The control line 112 operates to mask each column of scan cells, rather than masking an entire scan chain. Thus, individual bits from any scan chain can be masked on a per clock-cycle basis and the remaining bits of that scan chain

applied to the compactor 104. With control line 112 activated, all bits from the scan chains pass to the compactor. With control line 112 deactivated, all bits from the scan chains are masked. Although FIG. 10 shows only a single control line, additional control lines can be used to mask different groups of scan chains. Additionally,
5 although control line 112 is shown as active high, it may be configured as active low.

FIG. 11 shows yet another embodiment of the selective compactor 120. Automated test equipment 122 provides test patterns to the scan chains 124. The scan chains 124 are a part of the circuit under test 126. The patterns that are loaded into the scan chains 124 by the ATE are used to detect faults in the core logic of the circuit 126.
10 The test responses are stored in the scan chains 124 and are clocked in serial fashion to the selective compactor 120. The selective compactor includes a selector circuit 128 and a compactor 130. The selector circuit 128 includes control logic including an input register 132, multiple control registers 134, 136, and multiple decoders 137 and 139. The register 132 is loaded with a pattern of bits that are moved to the control registers
15 134, 136 upon activation of an update line (not shown). The control registers 134, 136 are read by the decoders 137 and 139 and decoded to select one or more logic gates 138. A flag 140 is used to override the decoders 137 and 139 and pass all of the test responses to the compactor 130. Although only a single flag 140 is shown, multiple flags may be used to separately control the decoders. In this example, the compactor
20 130 includes multiple spatial compactors, such as compactors 142 and 144. Each control register may be loaded with different data so that the compactors 142, 144 can be controlled independently of each other.

FIG. 12 shows yet another embodiment of the present invention with a selective compactor 150. Control logic 152 variably controls which test responses are masked
25 and which test responses are compacted. Thus, activating the corresponding bit position in the control logic 152 activates the corresponding logic gate associated with that bit and allows the test response to pass to the compactor. Conversely, any bit that is not activated masks the corresponding test response.

FIG. 13 shows another embodiment of a selective compactor 156 including a selector circuit 158 and compactor 160. In this case, an input shift register 162 having a bit position corresponding to each scan chain 164 is used to selectively mask the scan chains. A clock is applied to line 166 to serially move data applied on data line 168 into the shift register 162. At the appropriate time, an update line 165 is activated to move the data from the shift register to a control register 169. Each bit position that is activated in the control register 169 allows a test response from the scan chains 164 to pass to the compactor. All other test responses are masked. Thus, the selective compactor can mask any variable number of test responses.

Each of the embodiments described above can be used as a diagnostic tool for localizing faults in the circuit under test. For example, each test response can be analyzed individually by masking all other test responses in the scan chains connected to the same compactor. By viewing the test response individually, the bit position in the test response containing fault effects can be determined.

FIG. 14 shows a flowchart of a method for selectively compacting test responses. In process block 170, an ATE loads predetermined test patterns into scan chains within an integrated circuit. This loading is typically accomplished by shifting the test patterns serially into the scan chains. The test patterns are applied to the circuit under test (process block 172) and the test responses are stored in the scan chains (process block 174). In process block 176, the selector circuit controls which test responses are masked. In particular, the selector circuit controls which scan chains are masked or which bits in the scan chains are masked. For example, in FIG. 8, the selector circuit masks the entire scan chain that is identified in the scan identification field 58. In FIG. 10, only individual bits of a scan chain are masked. In any event, in process block 176, the selector circuit typically masks unknown data or multiple fault effects so that the desired fault effect can propagate to the output (in some modes of operation, all of the test responses may pass to the output). In the event that the selector circuit includes a control register, the control register may be loaded concurrently with loading the test patterns in the scan chains or it can be loaded prior to reading the test responses. In

process block 178, the test responses (one or more of which have been masked) are passed to the compactor and the compactor generates a compressed output associated with the test responses. In process block 180, the compressed output generated by the compactor is compared to an ideal response. If they match, the integrated circuit is
5 assumed to be fault free.

Having illustrated and described the principles of the illustrated embodiments, it will be apparent to those skilled in the art that the embodiments can be modified in arrangement and detail without departing from such principles. For example, any of the illustrated compactors can be used with any of the illustrated selector circuits with
10 minimum modification to create a selective compactor. Additionally, the selector circuit can easily be modified using different logic gates to achieve the selection functionality. For example, although the update lines are shown coupled to a separate bank of flip flops, the update lines can instead be coupled to input registers having tri-state outputs for controlling the logic in the selector circuit. Still further, although the scan chains are
15 shown as serial shift registers, logic may be added so as to output test response data in parallel to the selective compactor. Additionally, although multiple spatial and time compactors were shown, compactors having features of both spatial and time compactors may be used. Indeed, any conventional or newly developed compactor may be used with the selection circuitry.

20 In view of the many possible embodiments, it will be recognized that the illustrated embodiments include only examples of the invention and should not be taken as a limitation on the scope of the invention. Rather, the invention is defined by the following claims. We therefore claim as the invention all such embodiments that come within the scope of these claims.

25

We claim:

1. An apparatus for selectively compacting test responses of an integrated circuit, comprising:
 - a plurality of scan chains within the integrated circuit to store test responses that
 - 5 indicate faults in the integrated circuit;
 - at least one compactor to compress the test responses stored within the scan chains; and
 - a selector circuit coupled between the scan chains and the compactor that masks one or more of the test responses in the scan chains from the compactor.
- 10 2. The apparatus of claim 1, wherein the at least one compactor is selected from a group consisting of a spatial compactor, a time compactor, and combinations thereof.
- 15 3. The apparatus of claim 1, wherein the selector circuit includes a register and a decoder coupled to the register, the decoder being responsive to the register for controlling whether the test responses from the scan chains pass to the compactor.
- 20 4. The apparatus of claim 1, wherein the selector circuit includes a register that stores an identification of one of the scan chains and a flag that when activated allows only the test response associated with a scan chain identified in the register to pass to the compactor and that when deactivated allows all of the test responses pass to the compactor, except for the test response identified in the register.
- 25 5. The apparatus of claim 1, wherein each scan chain has an output and the selector circuit includes a set of logic gates having first inputs coupled to the outputs of the scan chains, second inputs coupled to one or more control lines, and outputs coupled to the compactor, wherein with the one or more control lines activated the logic gates pass data on the outputs of the scan chains to the compactor and with the one or more

control lines deactivated the logic gates prevent the data on the outputs of the scan chains from passing to the compactor.

5 6. The apparatus of claim 1 wherein each scan chain has an output and the selector circuit includes a set of logic gates having first inputs coupled to the outputs of the scan chains and second inputs coupled to a register, wherein data loaded into the register controls which scan chains are passed to the compactor.

10 7. The apparatus of claim 1 wherein the compactor and selector circuit are physically located on the integrated circuit.

8. The apparatus of claim 1 wherein the compactor and selector circuit are physically located externally of the integrated circuit.

15 9. The apparatus of claim 1 wherein masking a test response includes masking one or more bits of a test response in a scan chain.

10 10. The apparatus of claim 1 wherein masking a test response includes masking all bits in a scan chain that includes the test response.

20 11. A method for selectively compacting test responses of an integrated circuit, comprising:
passing N test responses in an integrated circuit to a selector circuit;
using the selector circuit, selectively preventing between 0 and N of the test
25 responses from being passed to a compactor while allowing the remaining test responses to be passed to the compactor; and
compacting the test responses passed to the compactor by the selector circuit.

12. The method of claim 11 wherein passing the test responses to the selector circuit includes shifting each test response serially to the selector circuit.

13. The method of claim 11 further including:

5 loading scan chains located within the integrated circuit with predetermined test patterns for testing the integrated circuit; and
generating and storing test responses in response to applying the test patterns to the integrated circuit.

10 14. The method of claim 11 further including:

loading scan chains located within the integrated circuit with predetermined test patterns;

generating and storing test responses in the scan chains;

identifying one of the scan chains; and

15 wherein selectively preventing includes preventing only the test response in the identified scan chain from being passed to the compactor.

15. The method of claim 11 further including:

20 loading scan chains located within the integrated circuit with predetermined test patterns;

storing test responses in the scan chains;

identifying one scan chain as a scan chain of interest and passing only the test response within the identified scan chain to the compactor while preventing all other test responses from being passed to the compactor.

25

16. The method of claim 11 further including loading a scan chain identifier into a register, the scan chain identifier uniquely identifying one of the scan chains.

17. The method of claim 11 wherein selectively preventing the test responses from being passed to the compactor includes preventing one or more bits of a test response from being passed to the compactor while allowing other bits of the test response to pass to the compactor.

5

18. The method of claim 11 where selectively preventing the test responses from being passed to the compactor includes preventing all bits of a test response from being passed to the compactor.

10

19. The method of claim 11 further including:

loading a plurality of scan chains located within the integrated circuit with test patterns;

storing the test responses in the scan chains;

15

loading a register with a scan chain identifier that uniquely identifies one of the scan chains; and

optionally configuring the selector circuit so that either all of the test responses in the scan chains are passed to the compactor except the test response stored in the scan chain identified in the register or only the test response stored in the scan chain identified in the register is passed to the compactor and all other test responses are masked from the compactor.

20

20. The method of claim 11 wherein the selectively preventing includes allowing all of the test responses to pass to the compactor.

25

21. An apparatus for selectively compacting test responses of an integrated circuit, comprising:

means for storing a plurality of test responses that include information relating to faults within the integrated circuit;

selection means for selectively masking the test responses; and

means for compacting the test responses selected by the selection means.

22. The apparatus of claim 21 wherein the means for selectively masking includes means for either allowing all of the test responses to pass to the means for
5 compacting or for masking any variable number of test responses to prevent the masked test responses from being passed to the means for compacting.

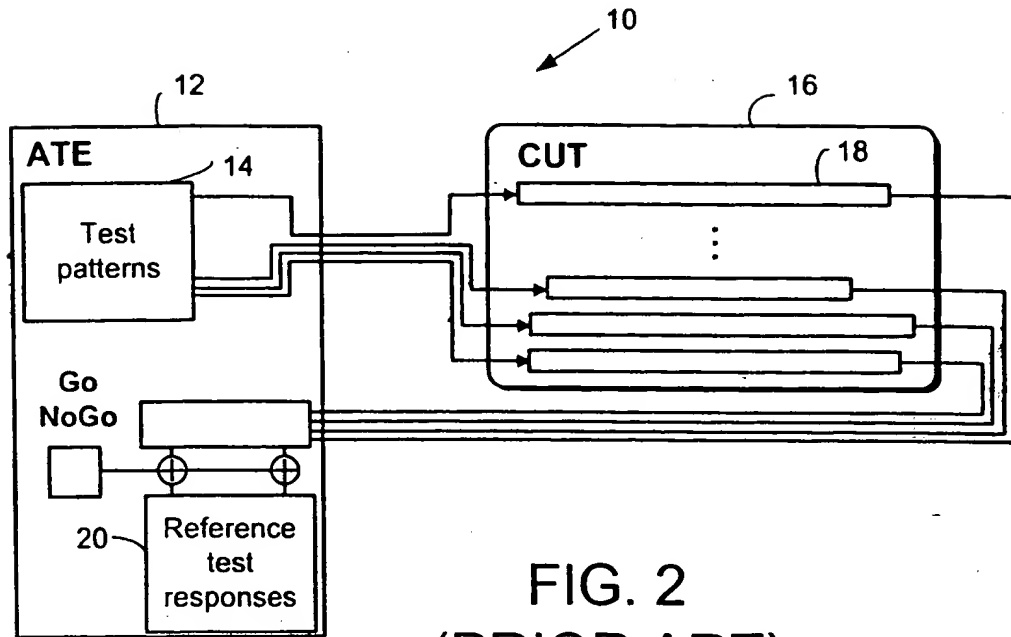
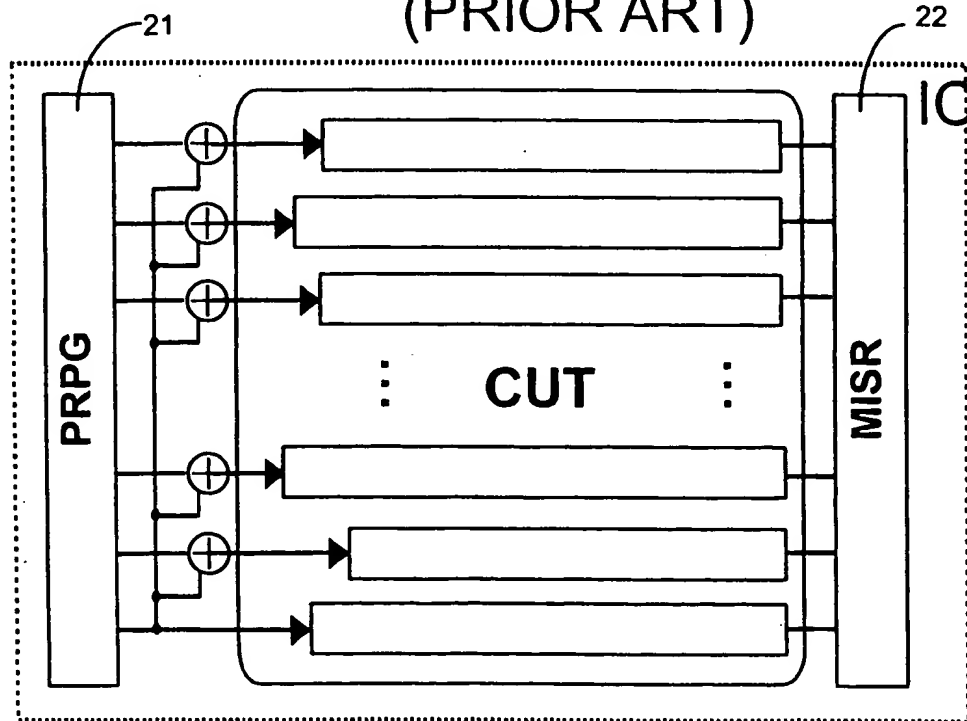
23. The apparatus of claim 21 wherein the means for selectively masking includes means for selecting between the following modes of operation:
10 a) all test responses are compacted; or
b) all test responses are compacted except one identified test response.

24. The apparatus of claim 21 wherein the means for selectively masking includes means for selecting between the following modes of operation:
15 a) all test responses are compacted; or
b) only one identified test response is compacted.

25. The apparatus of claim 21 wherein the means for storing a plurality of test responses includes scan chain means for serially receiving predetermined test patterns,
20 for applying the test patterns to the integrated circuit and for storing the test responses.

26. The apparatus of claim 21 wherein means for selectively masking includes decoding means for decoding a scan chain identification register.

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FIG. 1
(PRIOR ART)FIG. 2
(PRIOR ART)

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FIG. 3
(PRIOR ART)

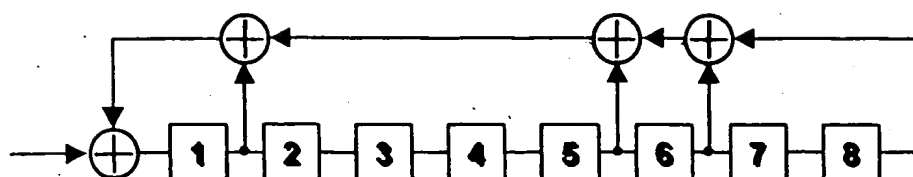


FIG. 4
(PRIOR ART)

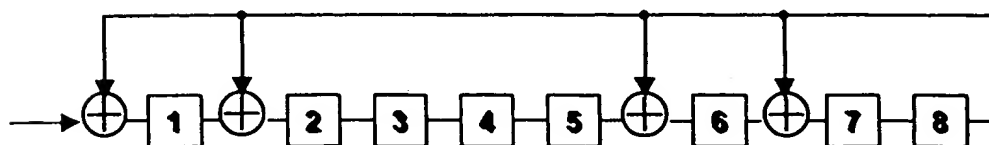
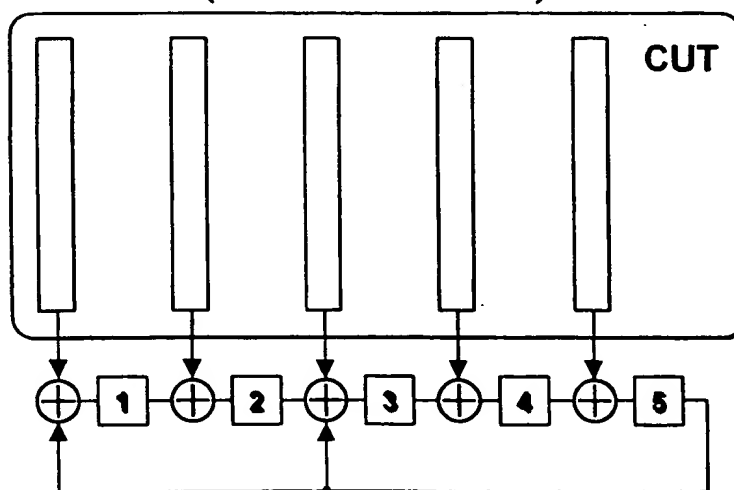
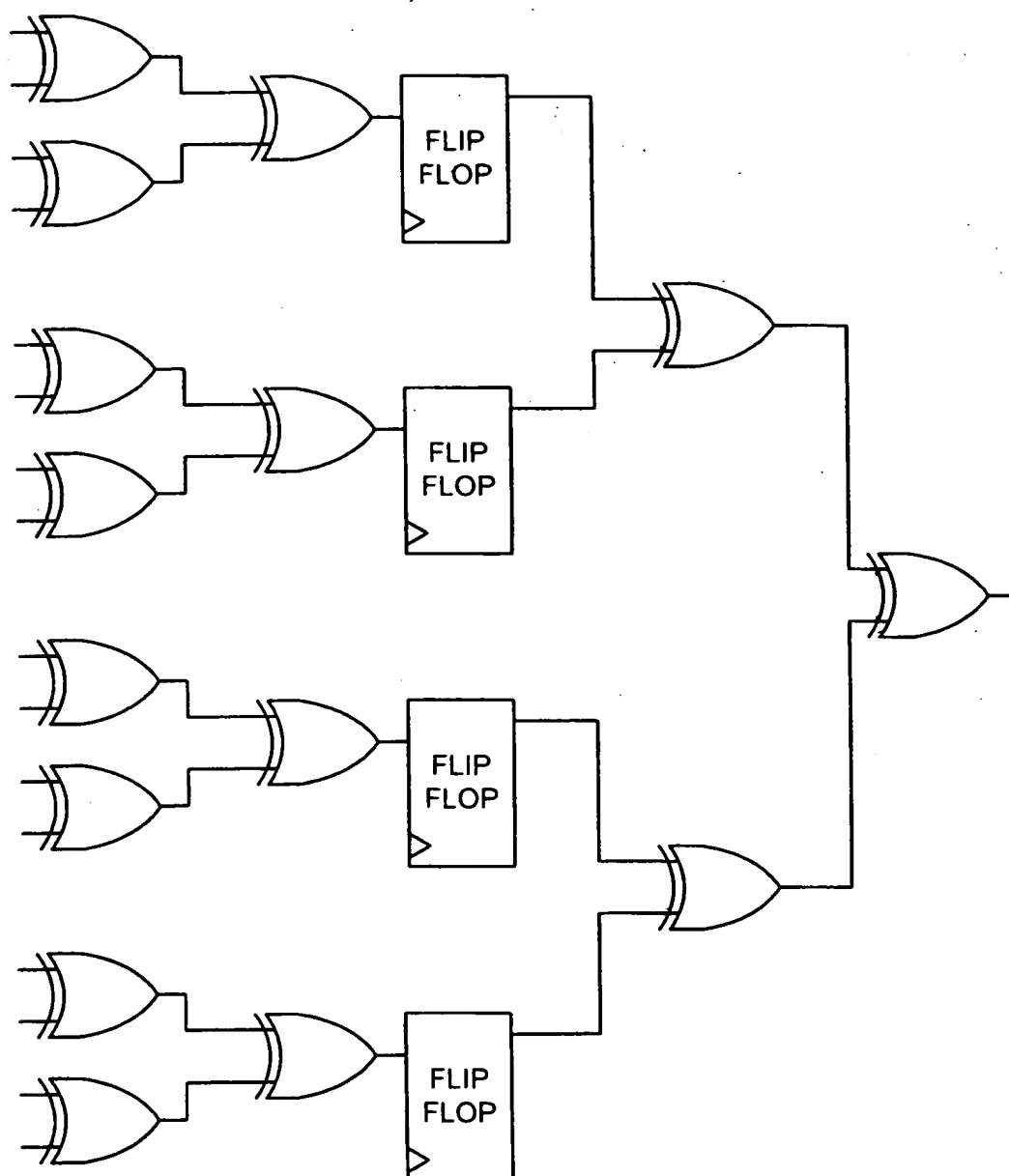


FIG. 5
(PRIOR ART)



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FIG. 6
(PRIOR ART)

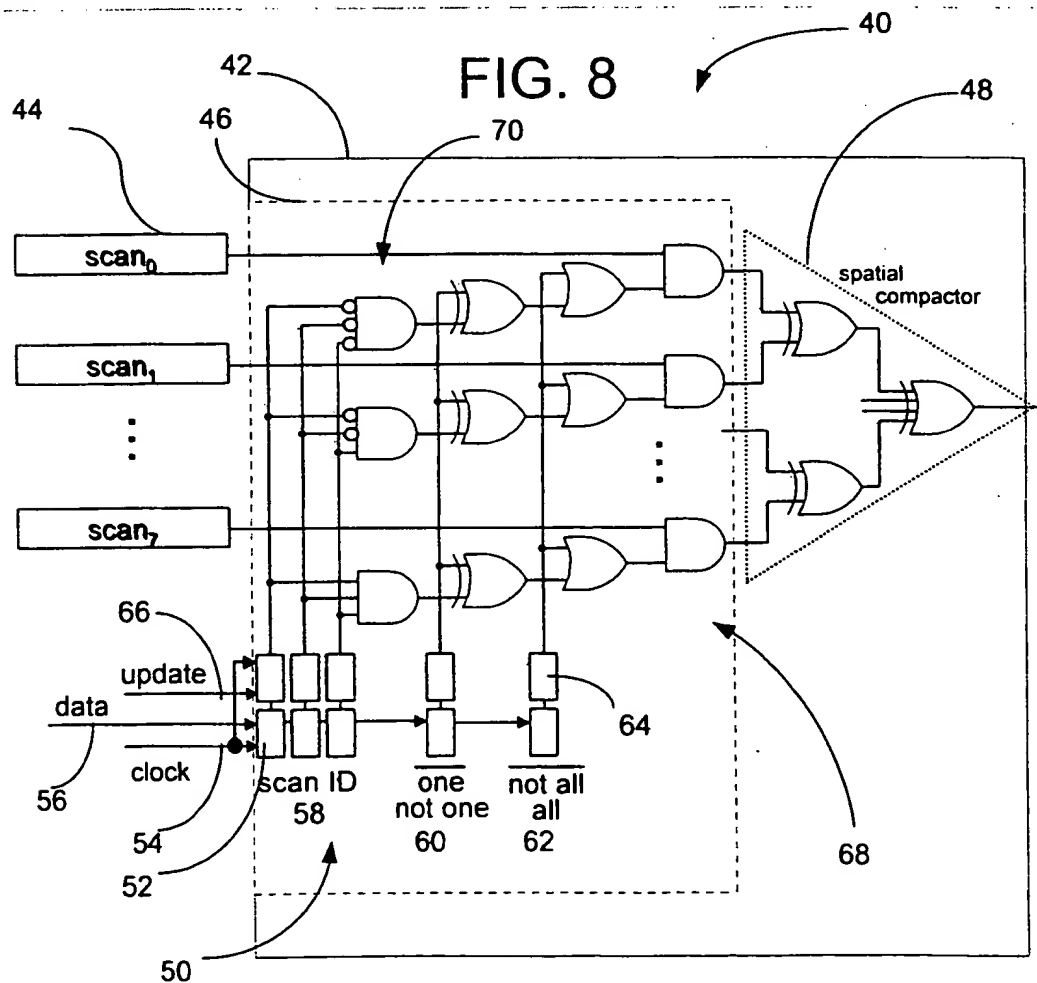


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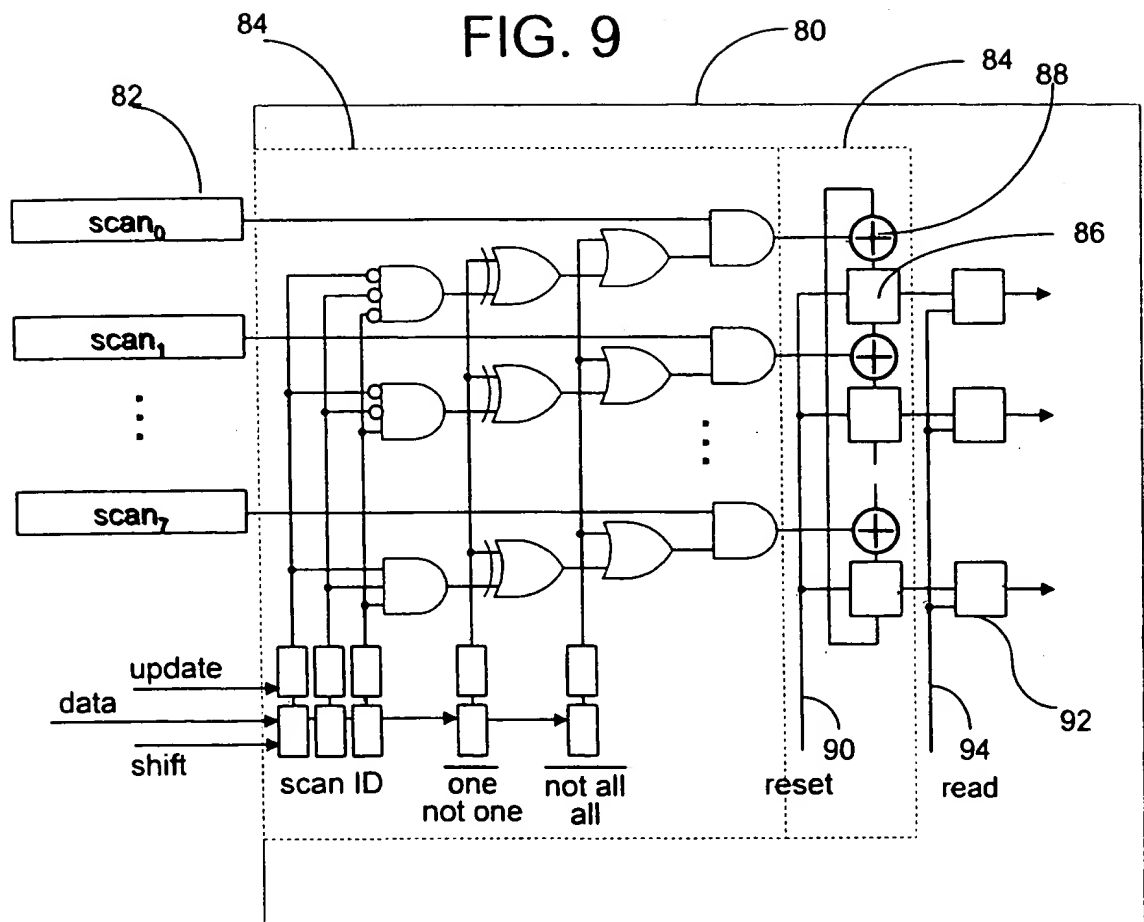
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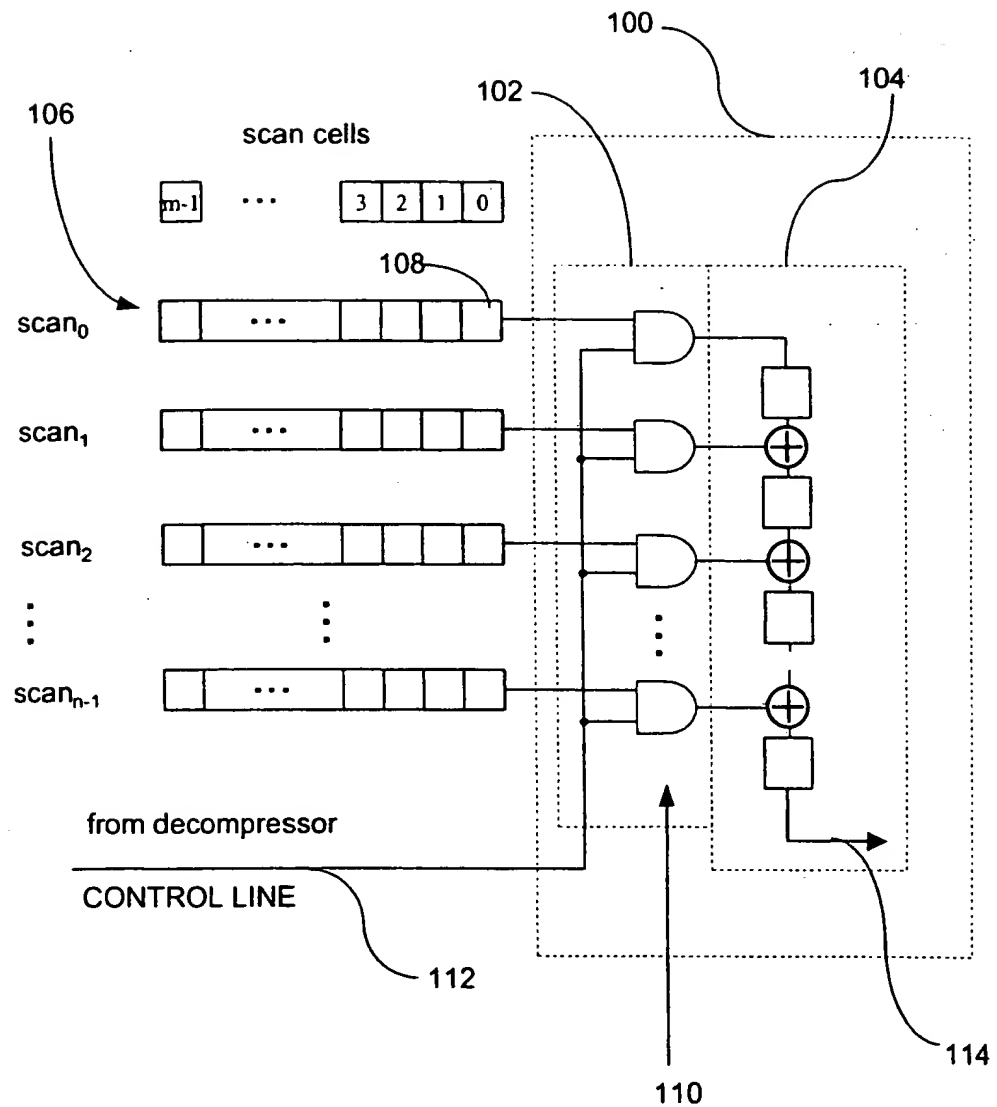


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FIG. 10



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FIG. 11

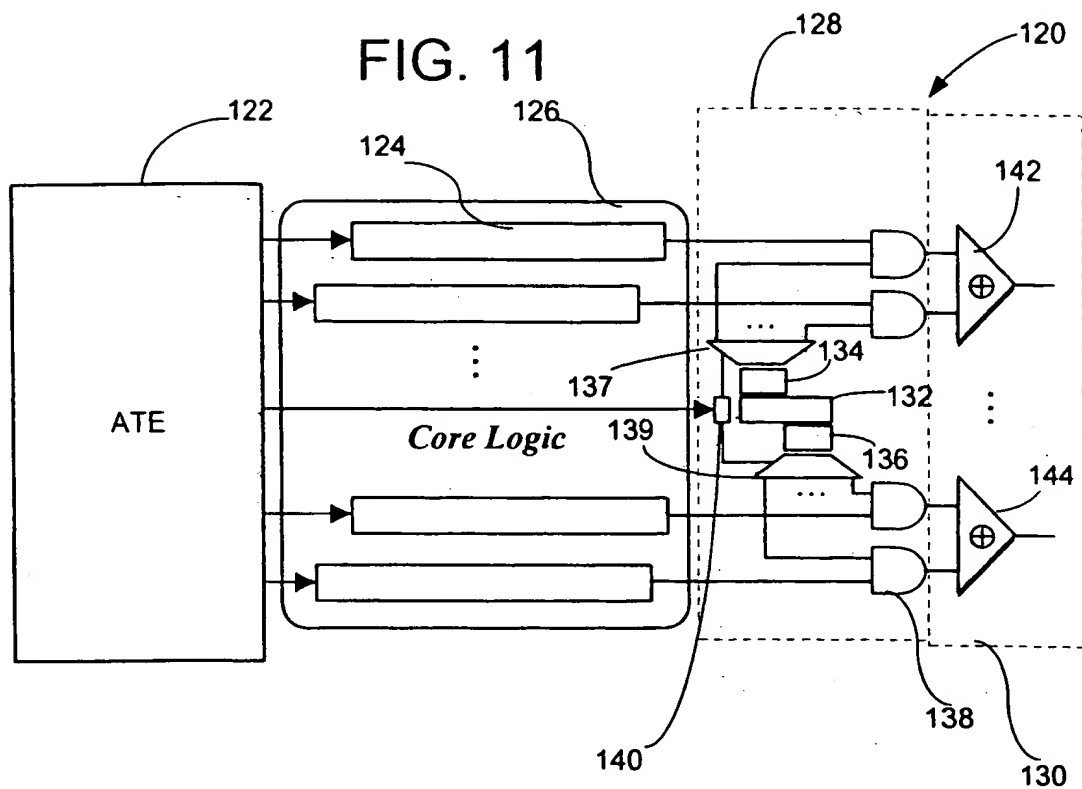
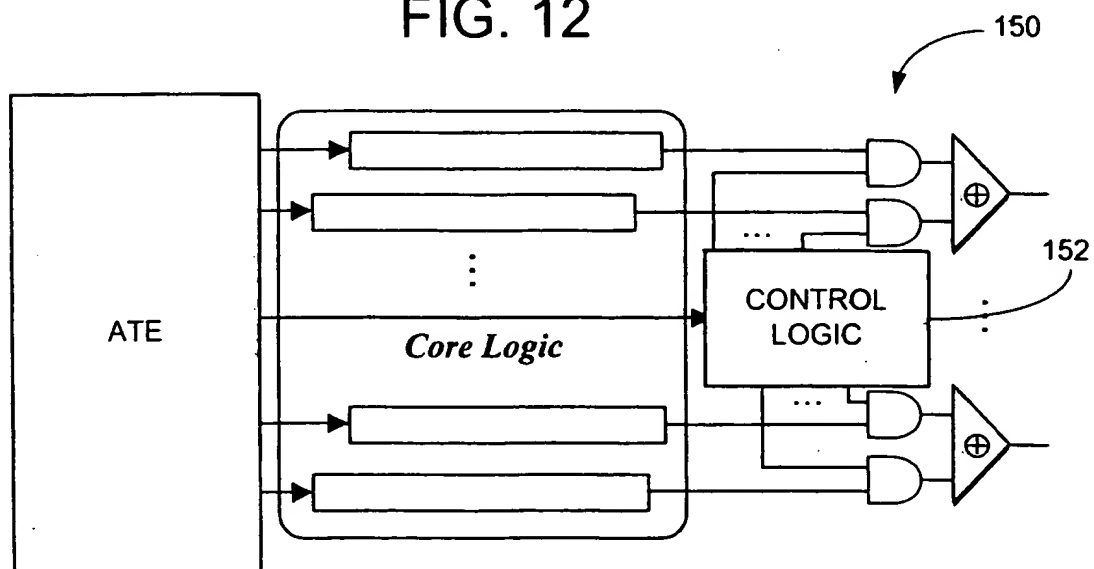
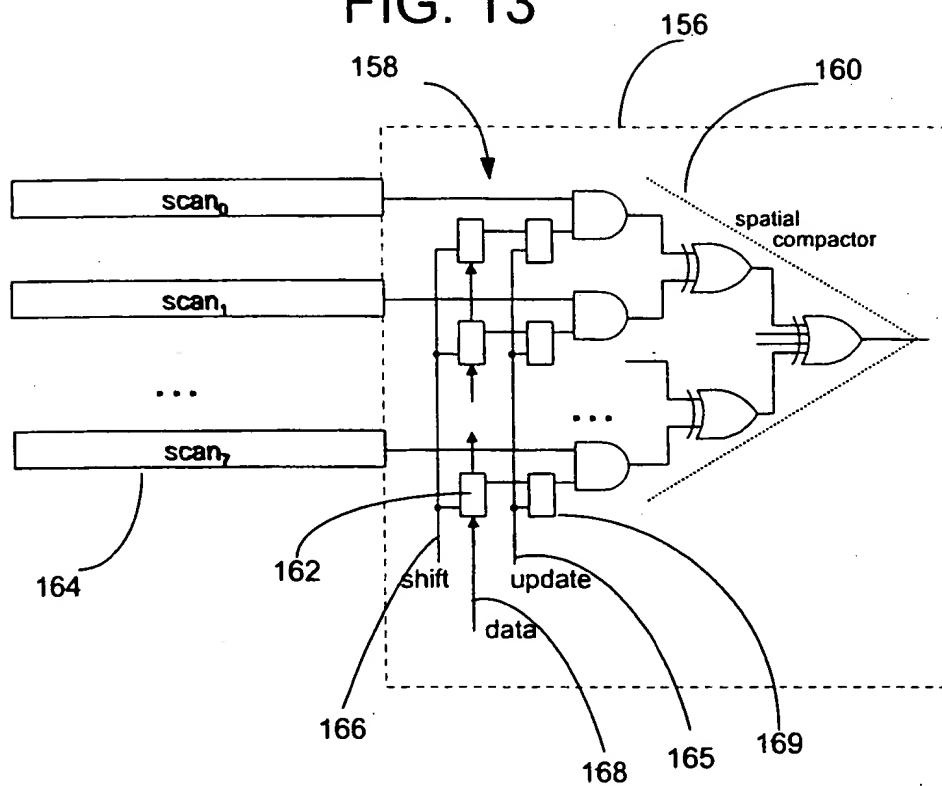


FIG. 12



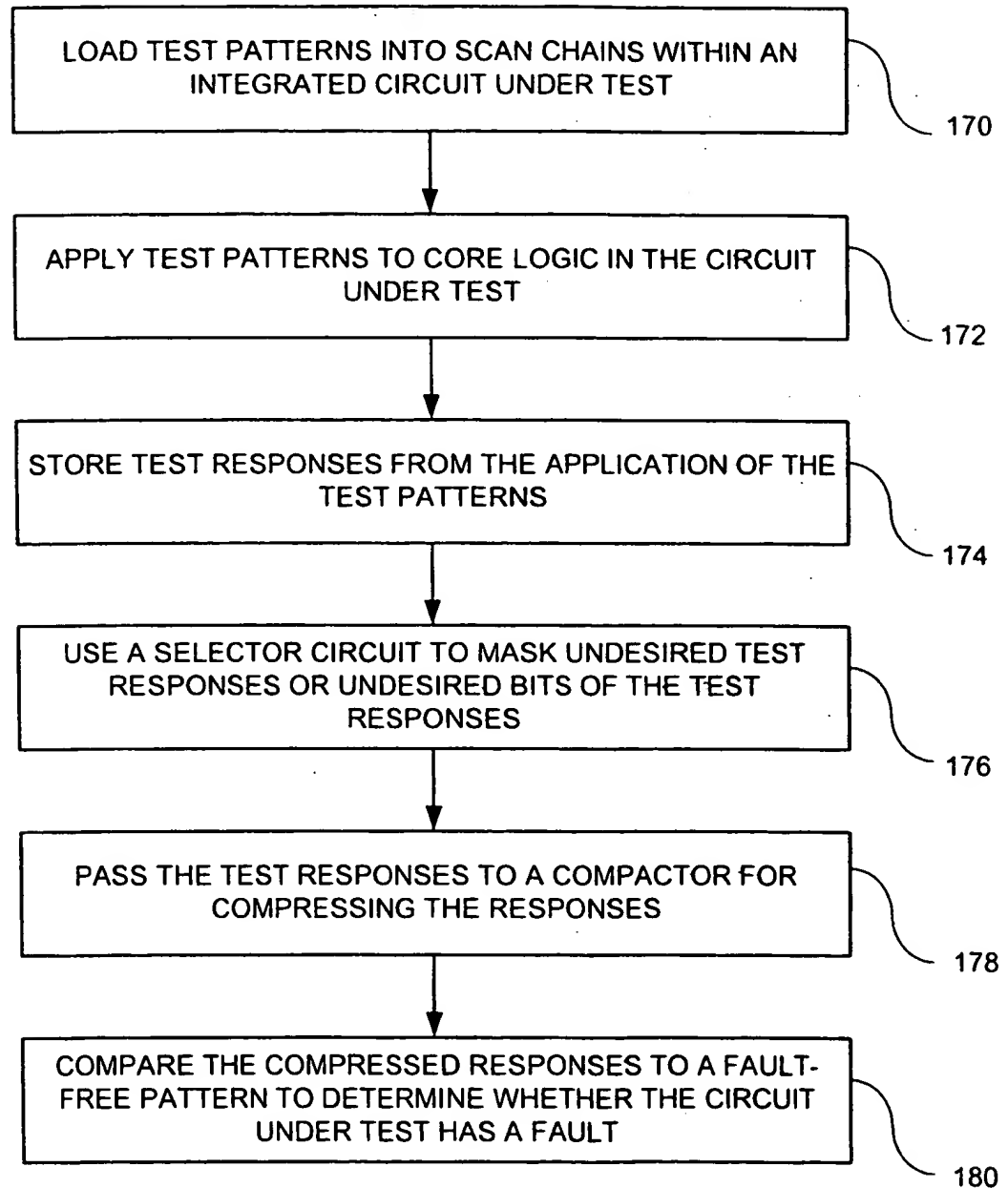
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FIG. 13



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FIG. 14



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INTERNATIONAL SEARCH REPORT

International application No.
PCT/US00/31376

A. CLASSIFICATION OF SUBJECT MATTER

IPC(7) : G01R 31/28

US CL : 714/729

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

U.S. : 714/729, 726, 738, 742, 712, 718, 820, 821, 30, 37, 43, 45; 370/241, 242

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practicable, search terms used)

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
Y	US 4,754,215A (KAWAI) 28 June 1998, col. 2 lines 18-58.	1-26
Y	US 5,790,562A (MURRAY al) 04 August 1998, col. 2 line 55-col. 7 line 30.	1-26
A	US 4,503,537A (MCANNEY) 05 March 1985, col. 2 lines 38-64.	1-26
A	US 4,503,537A (MCANNEY) 05 March 1985, col. 2 lines 38-64.	1-26

☐ Further documents are listed in the continuation of Box C. ☐ See patent family annex.

* Special categories of cited documents:	*T* later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention
A document defining the general state of the art which is not considered to be of particular relevance	*X* document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone
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L document which may throw doubts on priority claim(s) or which is cited to establish the publication date of another citation or other special reason (as specified)	*Z* document member of the same patent family
O document referring to an oral disclosure, use, exhibition or other means	
P document published prior to the international filing date but later than the priority date claimed	

Date of the actual completion of the international search

04 JANUARY 2001

Date of mailing of the international search report

26 FEB 2001

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